

30V 3A Synchronous Step-Down Regulator

Features

- Wide 4.5V to 30V Operating Input Range
- 3A Continuous Output Current
- No Schottky Diode Required
- 600KHz Frequency Operation
- Built-in Over Current Limit
- Built-in Over Voltage Protection
- Internal Soft start
- Integrated internal compensation
- Output Adjustable from 0.925V
- Short Protection with Hiccup-Mode
- Thermal Shutdown
- Available in ESOP8 Package
- -40°C to +85°C Temperature Range

Applications

- Car Charge ports
- Smart power-meter system
- General purpose with high voltage input

General Description

The TX9826 is a high frequency, synchronous, rectified, step-down, switch-mode converter with internal power MOSFETs. It offers a very compact solution to achieve a 3A output current over a wide input supply range, with excellent load and line regulation.

The TX9826 requires a minimal number of readily available, external components and is available in a space saving ESOP8 package.

Typical Application Circuit

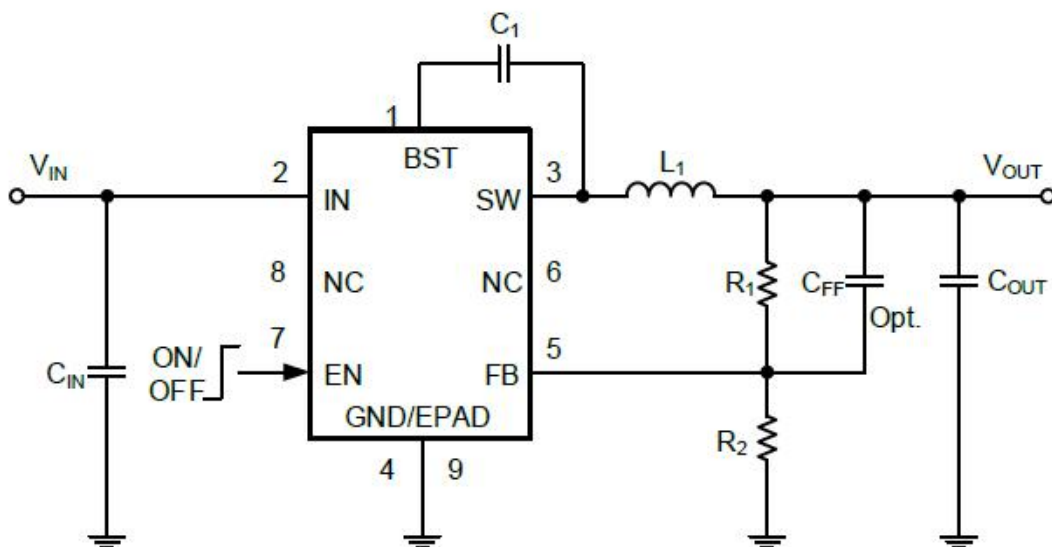


Figure 1. Basic Application Circuit

Absolute Maximum Ratings

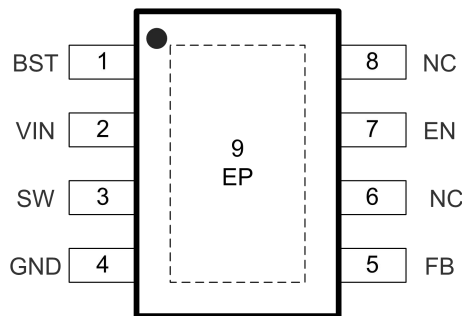
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Vin, EN, Voltage -0.3V to 36V
 Operating Temperature Range -40°C to +85°C
 FB Voltages -0.3 to 6V
 Lead Temperature (Soldering, 10s) +300°C
 Thermal Resistance (θJA) 50 °C/W

Thermal Resistance (θJC) 10 °C/W
 SW Voltage -0.3V to (VIN+0.5V)
 Storage Temperature Range -65°C to 150°C
 BS Voltage (Vsw-0.3) to (Vsw+5V)
 ESD (Machine Made) MM 200V
 ESD(Human Body Made) HMB 2KV

Pin Description

Pin Configuration



(ESOP8)

Top Marking: GMYLL (device code: GM, Y=year code, LL=lot number code)

Pin Description

Pin	Name	Function
1	BST	Bootstrap. A capacitor connected between SW and BST pins is required to form a floating supply across the high-side switch driver.
2	VIN	Power Supply Pin
3	SW	Switching Pin
4	GND	GROUND Pin
5	FB	Adjustable Version Feedback input. Connect FB to the center point of the external resistor divider
6	NC	
7	EN	Drive this pin to a logic-high to enable the IC. Drive to a logic-low to disable the IC and enter micro-power shutdown mode.
8	NC	
9	EPAD	GROUND Pin

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Electrical Characteristics

Parameter	Conditions	Min	Typ.	Max	Unit
Input Voltage Range		4.5		30	V
Supply Current in Operation	$V_{EN}=3.0V, V_{FB}=1.1V$		0.4	0.6	mA
Supply Current in Shutdown	$V_{EN}=0$ or $EN = GND$		4		μA
Regulated Feedback Voltage	$T_A = 25^\circ C, 4.5V \leq V_{IN} \leq 30V$	0.904	0.923	0.942	V
High-Side Switch On-Resistance			100		m Ω
Low-Side Switch On-Resistance			70		m Ω
High-Side Switch Leakage Current	$V_{EN}=0V, V_{SW}=0V$		0	10	μA
Upper Switch Current Limit	Minimum Duty Cycle		4		A
Oscillation Frequency			600		KHz
Maximum Duty Cycle	$V_{FB}=0.6V$		92		%
Minimum On-Time			60		nS
Minimum Off-Time			90		nS
Input OVP Voltage			31		V
Thermal Shutdown			160		$^\circ C$
Thermal Hysteresis			20		$^\circ C$

Functional Block Diagram

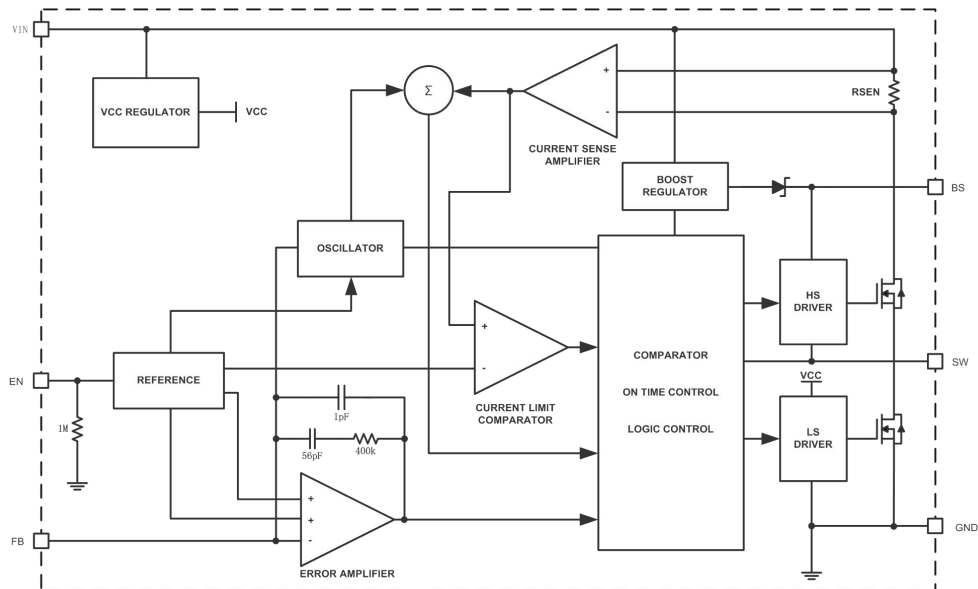


Figure 2. TX9826 Block Diagram

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Functions Description

Internal Regulator

The TX9826 is a current mode step down DC/DC converter that provides excellent transient response with no extra external compensation components. This device contains an internal, low resistance, high voltage power MOSFET, and operates at a high 600K operating frequency to ensure a compact, high efficiency design with excellent AC and DC performance.

Error Amplifier

The error amplifier compares the FB pin voltage with the internal FB reference (VFB) and outputs a current proportional to the difference between the two. This output current is then used to charge or discharge the internal compensation network to form the COMP voltage, which is used to control the power MOSFET current. The optimized internal compensation network minimizes the external component counts and simplifies the control loop design.

Internal Soft-Start

The soft-start is implemented to prevent the converter output voltage from overshooting during startup. When the chip starts, the internal circuitry generates a soft-start voltage (SS) ramping up from 0V to 0.923V. When it is lower than the internal reference (REF), SS overrides REF so the error amplifier uses SS as the reference. When SS is higher than REF, REF regains control. The SS time is internally fixed to 1.6ms.

Over Current Protection & Hiccup

has cycle-by-cycle over current limit when the inductor current peak value exceeds the set current limit threshold. Meanwhile, output voltage starts to drop until FB is below the Under-Voltage (UV) threshold, typically 25% below the reference. Once a UV is triggered, the Y3831 enters hiccup mode to periodically restart the part. This protection mode is especially useful when the output is dead-short to ground. The average short circuit current is greatly reduced to alleviate the thermal issue and to protect the regulator. The TX9826 exits the hiccup mode once the over current condition is removed.

Startup and Shutdown

If both VIN and EN are higher than their appropriate thresholds, the chip starts. The reference block starts first, generating stable reference voltage and currents, and then the internal regulator is enabled. The regulator provides stable supply for the remaining circuitries. Three events can shut down the chip: EN low, VIN low and thermal shutdown. In the shutdown procedure, the signaling path is first blocked to avoid any fault triggering. The COMP voltage and the internal supply rail are then pulled down. The floating driver is not subject to this shutdown command.

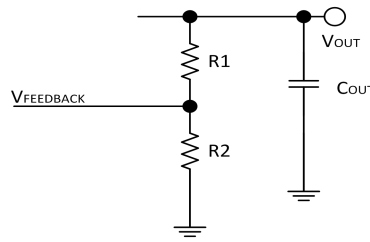
Applications Information

Setting the Output Voltage

TX9826 require an input capacitor, an output capacitor and an inductor. These components are critical to the performance of the device. TX9826 are internally compensated and do not require external components to achieve

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stable operation. The output voltage can be programmed by resistor divider.



Selecting the Inductor

The recommended inductor values are shown in the Application Diagram. It is important to guarantee the inductor core does not saturate during any foreseeable operational situation. The inductor should be rated to handle the peak load current plus the ripple current: Care should be taken when reviewing the different saturation current ratings that are specified by different manufacturers. Saturation current ratings are typically specified at 25°C, so ratings at maximum ambient temperature of the application should be requested from the manufacturer.

$$L = \frac{V_{out} \times (V_{in} - V_{out})}{V_{in} \times \Delta I_L \times f_{OSC}}$$

Where ΔI_L is the inductor ripple current. Choose inductor ripple current to be approximately 30% of the maximum load current, 3A. The maximum inductor peak current is:

$$I_{L(MAX)} = I_{LOAD} + \frac{\Delta I_L}{2}$$

Under light load conditions below 100mA, larger inductance is recommended for improved efficiency.

Selecting the Output Capacitor

Special attention should be paid when selecting these components. The DC bias of these capacitors can result in a capacitance value that falls below the minimum value given in the recommended capacitor specifications table.

The ceramic capacitor's actual capacitance can vary with temperature. The capacitor type X7R, which operates over a temperature range of -55°C to +125°C, will only vary the capacitance to within $\pm 15\%$. The capacitor type X5R has a similar tolerance over a reduced temperature range of -55°C to +85°C. Many large value ceramic capacitors, larger than 1 μ F are manufactured with Z5U or Y5V temperature characteristics. Their capacitance can drop by more than 50% as the temperature varies from 25°C to 85°C. Therefore X5R or X7R is recommended over Z5U and Y5V in applications where the ambient temperature will change significantly above or below 25°C.

Tantalum capacitors are less desirable than ceramic for use as output capacitors because they are more expensive when comparing equivalent capacitance and voltage ratings in the 0.47 μ F to 44 μ F range. Another important consideration is that tantalum capacitors have higher ESR values than equivalent size ceramics. This means that while it may be possible to find a tantalum capacitor with an ESR value within the stable range, it would have to be larger in capacitance (which means bigger and more costly) than a ceramic capacitor with the same ESR value. It should also be noted that the ESR of a typical tantalum will increase about 2:1 as the temperature goes from 25°C down to -40°C, so some guard band must be allowed.

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PCB Layout Consideration

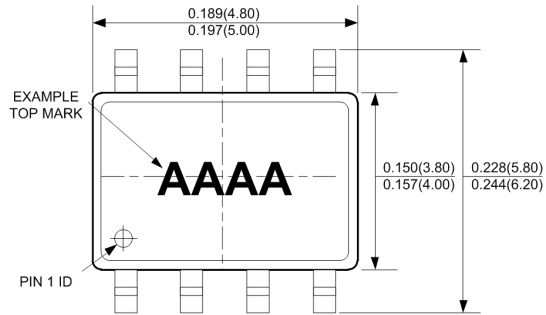
PCB layout is very important to achieve stable operation. It is highly recommended to duplicate EVB layout for optimum performance. If change is necessary, please follow these guidelines and take Figure 4 for reference.

1. Keep the path of switching current short and minimize the loop area formed by Input capacitor, high-side MOSFET and low-side MOSFET.
2. Bypass ceramic capacitors are suggested to be put close to the Vin Pin.
3. Ensure all feedback connections are short and direct. Place the feedback resistors and compensation components as close to the chip as possible.
4. VOUT, SW away from sensitive analog areas such as FB.
5. Connect IN, SW, and especially GND respectively to a large copper area to cool the chip to improve thermal performance and long-term reliability.
6. An example of 2-layer PCB layout is shown in Figure 4 for reference.

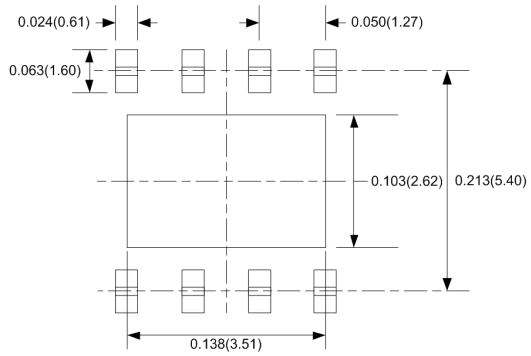
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Package Description

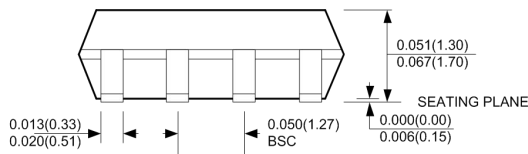
SOP8(EXPOSED PAD)



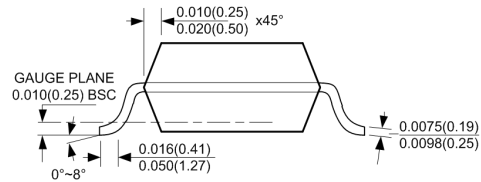
TOP VIEW



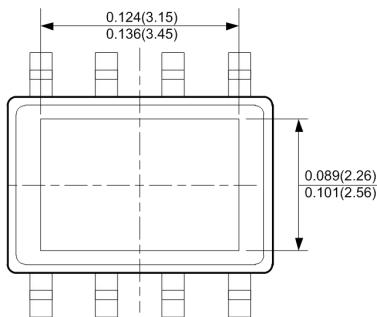
RECOMMENDED SOLDER PAD LAYOUT



FRONT VIEW



SIDE VIEW



BOTTOM VIEW

- NOTE:
- 1) CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.
 - 2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
 - 3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
 - 4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.004" INCHES MAX.
 - 5) DRAWING CONFORMS TO JEDEC MS-012, VARIATION BA.
 - 6) DRAWING IS NOT TO SCALE.